

02417: Time Series Analysis

# Week 7 – Linear systems

Peder Bacher  
DTU Compute

Based on material previous material from the course

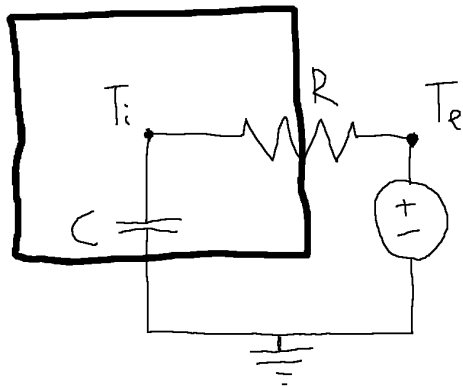
March 22, 2025

## Week 7: Outline of the lecture

- ▶ Input-Output systems, Sec. 4 introduction and 4.1
- ▶ Linear system notation
- ▶ The  $z$ -transform, Sec. 4.4
- ▶ Cross Correlation Functions – from Sec. 6.2.2
- ▶ Transfer function models; identification, estimation, validation, prediction, Chap. 8

# Simplest first order RC-system

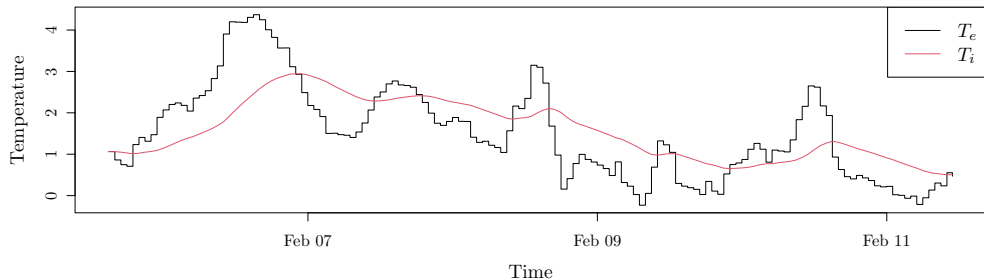
Single state model of the temperature in a box:



# Simplest RC-system

- ▶  $T_t^e$  external and  $T_t^i$  internal temperature at time  $t = [1, 2, \dots, n]$
- ▶ ODE model

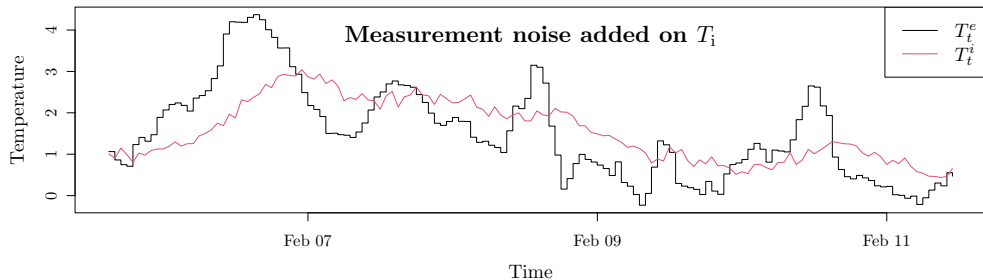
$$\frac{dT_i}{dt} = \frac{1}{RC}(T_e - T_i)$$



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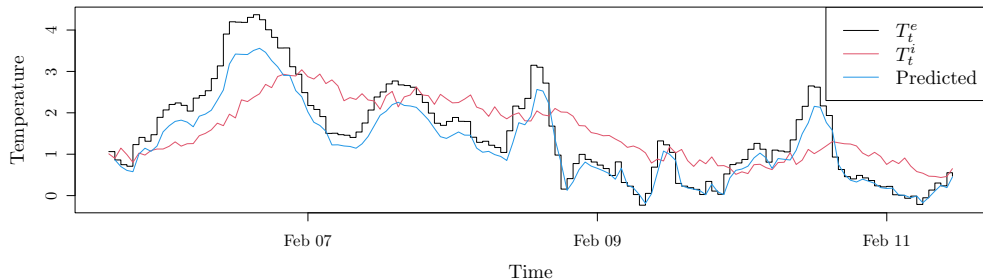


# Try a static model

- ▶ A simple linear regression model ( $\varepsilon_t$  is the error)

$$T_t^i = \omega_e T_t^e + \varepsilon_t$$

- ▶ Are the dynamics well described by the model?



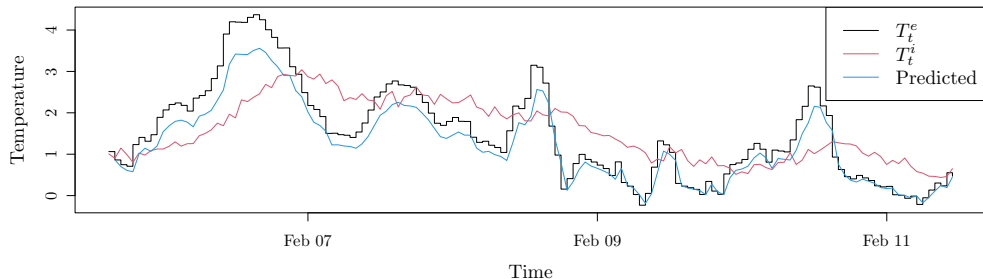
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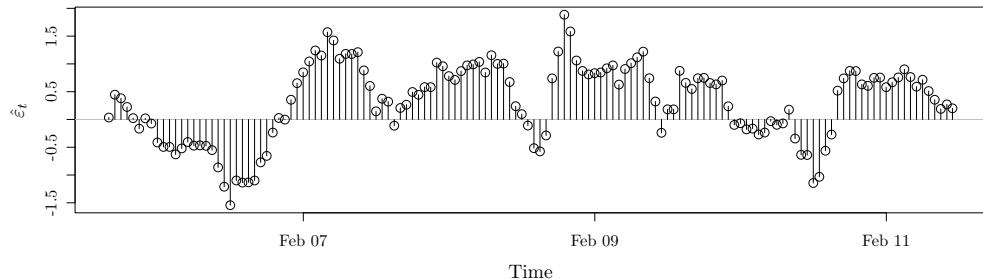
No, the predicted temperature is just proportional to the input ( $T_t^i$ )



# Model validation: check i.i.d. of residuals

Are residuals like white noise?

- ▶ Check if they are *independent and identically distributed*
- ▶ Is  $\hat{\varepsilon}_t$  independent of  $\hat{\varepsilon}_{t-k}$  for all  $t$  and  $k$ ?



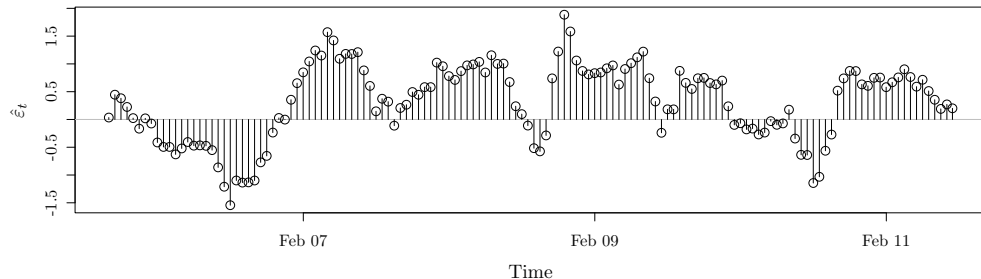


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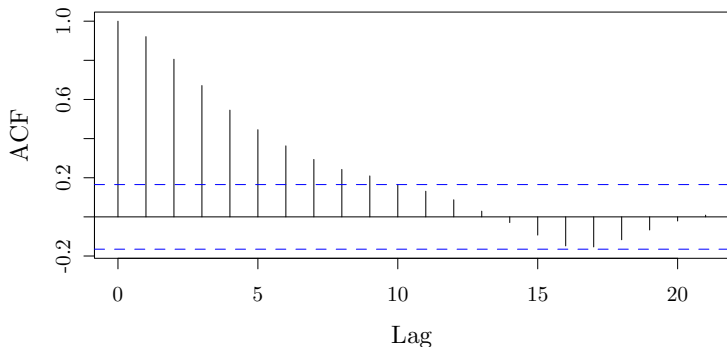
- ▶ Check if they are *independent and identically distributed*
- ▶ Is  $\hat{\varepsilon}_t$  independent of  $\hat{\varepsilon}_{t-k}$  for all  $t$  and  $k$ ?

Nope! There is a pattern left...



# Model validation: Test for i.i.d. with ACF

TEST if residuals are white noise?

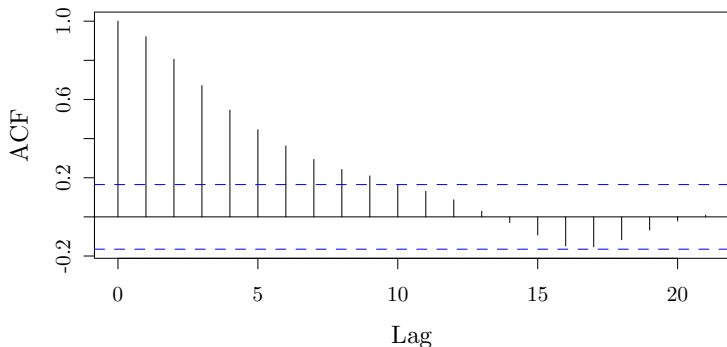


It's not white noise!

How do we find a better model?

# Model validation: Test for i.i.d. with ACF

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It's not white noise!

How do we find a better model? The exponential decay in ACF points to an AR part!

$$\frac{dT_i}{dt} = \frac{1}{RC}(T_e - T_i)$$

It has the solution

$$T_i(t + \Delta t) = T_e(t) + e^{-\frac{\Delta t}{RC}}(T_i(t) - T_e(t))$$

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$$T_{t+1}^i = e^{-\frac{1}{RC}} T_t^i + (1 - e^{-\frac{1}{RC}}) T_t^e$$

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since  $e^{-\frac{1}{RC}}$  is between 0 and 1, then write it as

$$T_{t+1}^i = \phi_1 T_t^i + \omega_1 T_t^e$$

where  $\phi_1$  and  $\omega_1$  are between 0 and 1.

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Add a noise term and we have the Auto-Regressive with eXogeneous input (ARX) model

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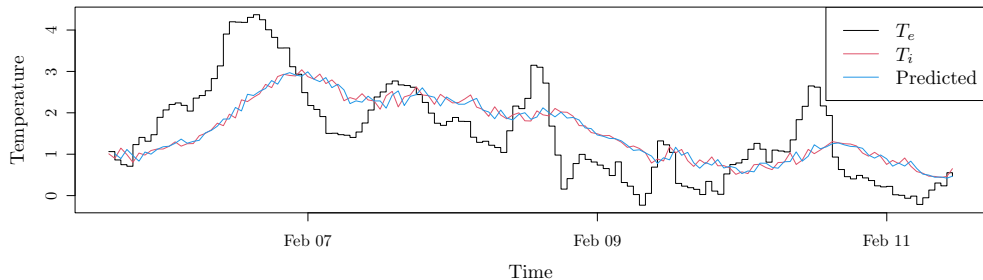
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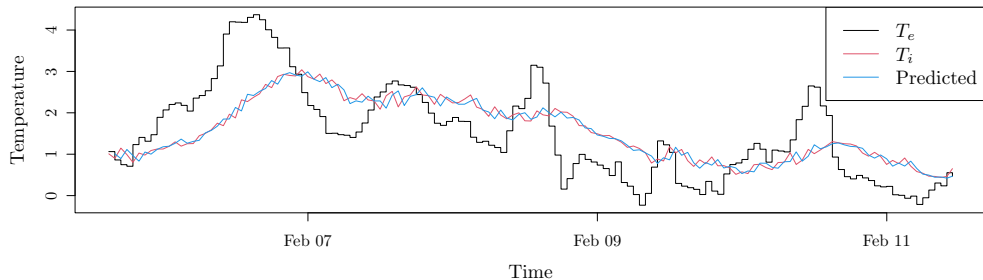
An ARX model

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## ARX model

## ► The residuals

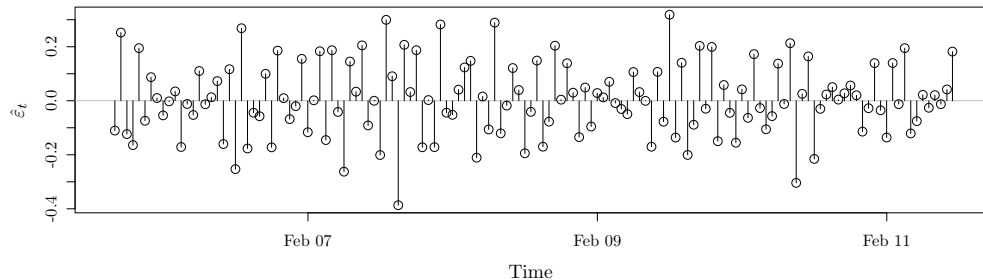
$$\hat{\varepsilon}_t = \text{obs.} - \text{pred.} = T_t^i - \hat{T}_t^i = T_t^i - \frac{\hat{\omega}_1 B}{1 - \hat{\phi}_1 B} T_t^e$$

## ARX model

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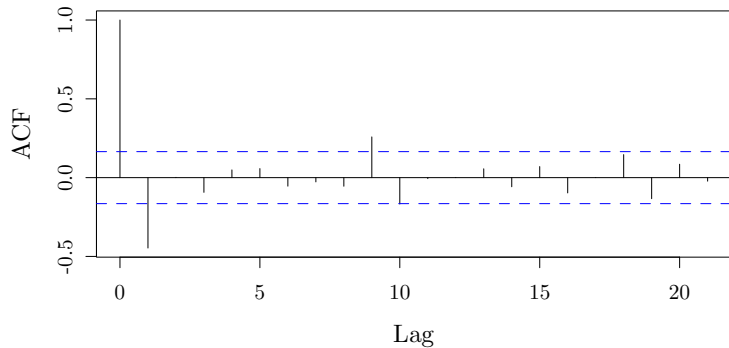
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- ▶ Are the residuals now white noise?



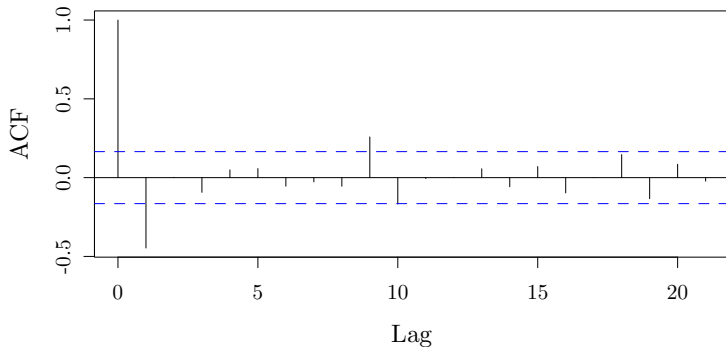
# Check for i.i.d. of residuals

Is it likely that this is white noise?



## Check for i.i.d. of residuals

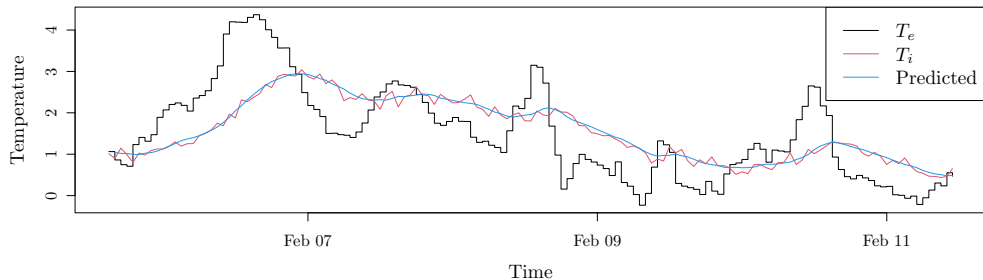
Is it likely that this is white noise? Almost!



Actually we miss an MA part!

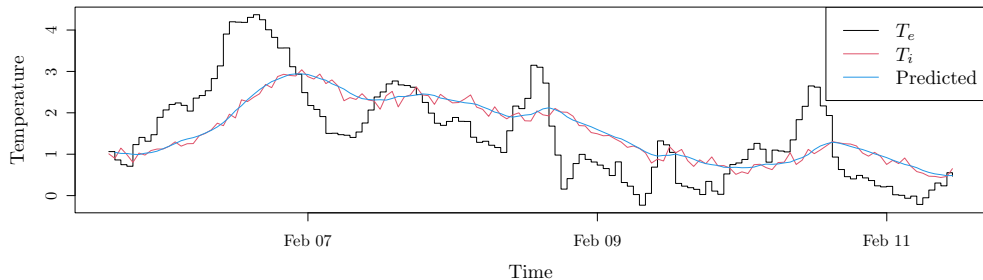
## An ARMAX model

$$T_t^i = \phi_1 T_{t-1}^i + \omega_1 T_t^e + \varepsilon_t + \theta_1 \varepsilon_{t-1}$$



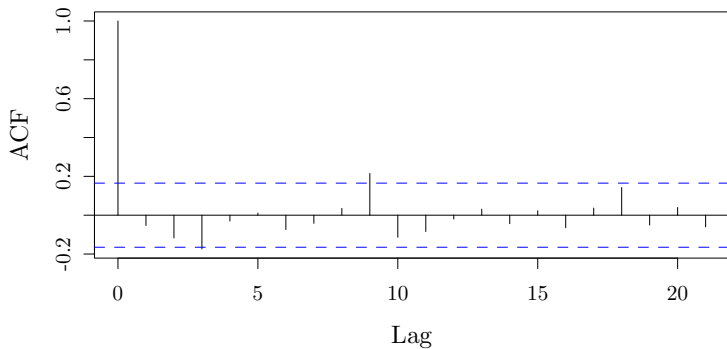
## An ARMAX model

$$Y_t = \phi_1 Y_{t-1} + \omega_1 x_t + \varepsilon_t + \theta_1 \varepsilon_{t-1}$$



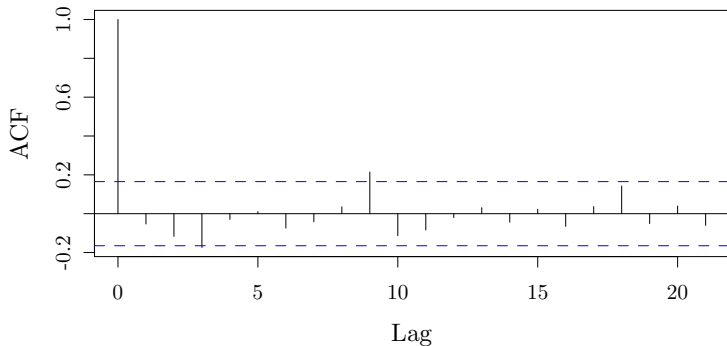


# Validate the model with the residuals ACF



Now we have *white noise residuals* :-)

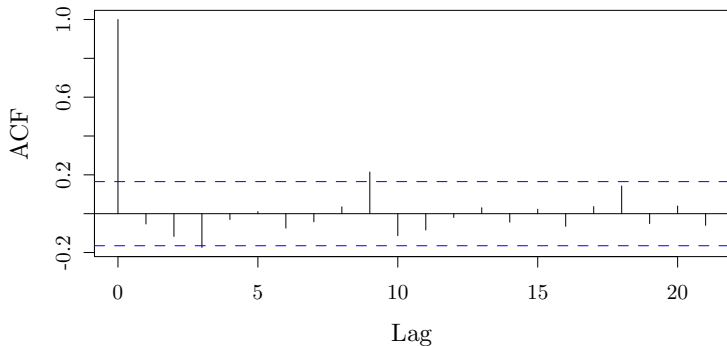
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Remember, we are validating the *one-step prediction* residuals:  $\hat{\epsilon}_{t+1} = y_{t+1} - \hat{y}_{t+1|t}$

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Now we have *white noise residuals* :-)

Remember, we are validating the *one-step prediction* residuals:  $\hat{\varepsilon}_t = y_t - \hat{y}_{t|t-1}$

## Dependence between variables: Cross-correlation function

Calculate the Cross-Correlation Function (CCF) by simply shifting the index and lag *another* series:

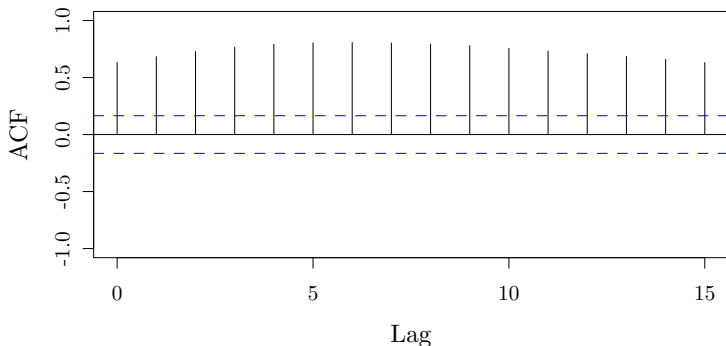
t	$Y_t$	$X_t$	$X_{t-1}$
1	4	2	
2	5	3	2
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4	3	3	8
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8			8

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Cross-Correlation Function (CCF) between input and output

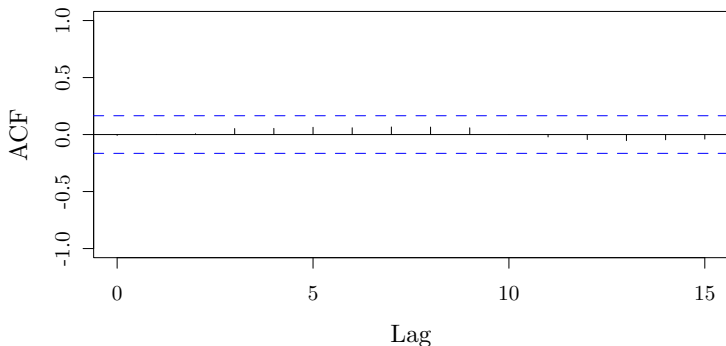


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Cross-Correlation Function (CCF) between input and residuals.

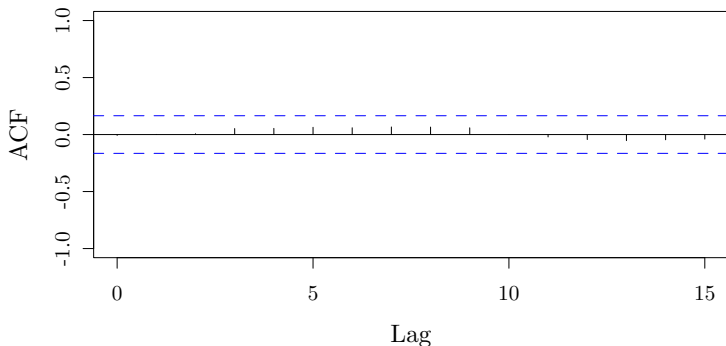


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Cross-Correlation Function (CCF) between input and residuals.



So we can **use the CCF in model validation**: If no correlation between input(s) and residuals, then the model is good!





# Dynamic Systems – Some characteristics

**Def. Linear system:**

$$\mathcal{F}[\lambda_1 x_1(t) + \lambda_2 x_2(t)] = \lambda_1 \mathcal{F}[x_1(t)] + \lambda_2 \mathcal{F}[x_2(t)]$$

**Def. Time invariant system:**

$$y(t) = \mathcal{F}[x(t)] \Rightarrow y(t - \tau) = \mathcal{F}[x(t - \tau)]$$

**Def. Stable system:** A system is said to be *stable* if any constrained input implies a constrained output.

**Def. Causal system:** A system is said to be *physically feasible* or *causal*, if the output at time  $t$  does not depend on future values of the input.

## Example: "ARX(1)" system

► *System:*  $y_t - ay_{t-1} = bx_t$

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Yes: Time invariant since the coefficients don't change in time



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▶ Is the system *causal*?

Yes:  $y_t$  depend only on past input values  $x_t, x_{t-1}, \dots$

▶ Is the system *stable*?

Yes, for  $|a| < 1$  the coefficients sum is bounded

$$\sum_{k=0}^{\infty} |a|^k = \begin{cases} 1/(1 - |a|) & ; |a| < 1 \\ \infty & ; |a| \geq 1 \end{cases}$$

(stability does not depend on  $b$ )

## Def. discrete impulse and step reponse

For *linear time invariant systems* the input can be convoluted to get the output:

- ▶ Discrete time:

$$y_t = \sum_{k=-\infty}^{\infty} h(k)x_{t-k} \quad (1)$$

Causal, then:

$$y_t = \sum_{k=0}^t h(k)x_{t-k} \quad (2)$$

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- ▶  $S_k = \sum_{j=-\infty}^k h_j$  is called the *step response*, why? What happens if  $x_k = 1$  for all  $k$ ?

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$$y_t = 0.8y_{t-1} + 2x_t$$

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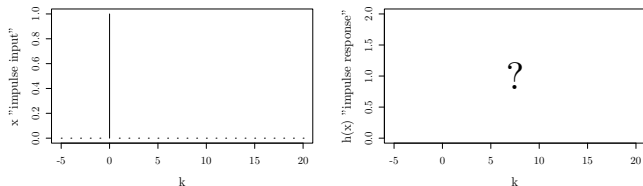
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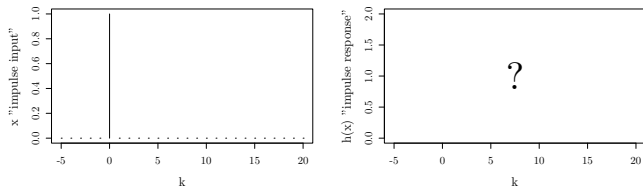
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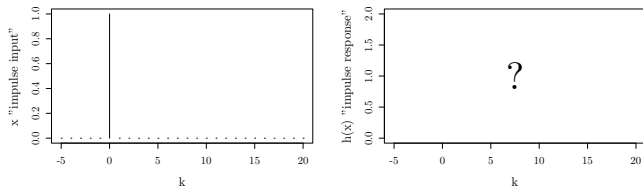
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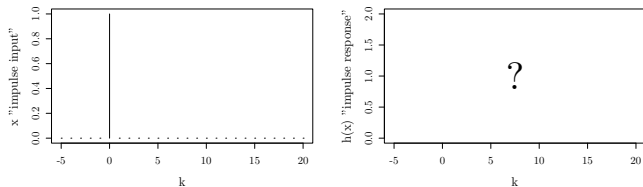
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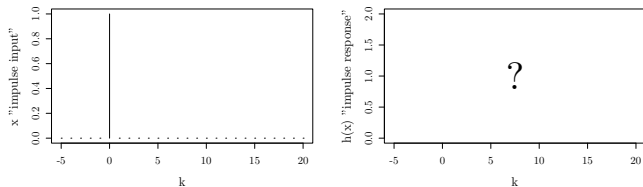
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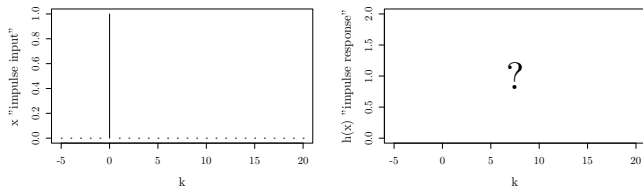
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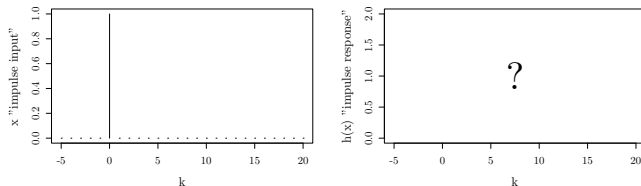
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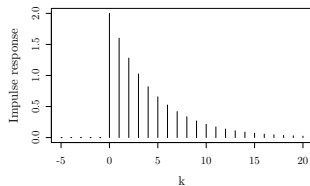
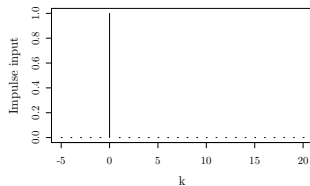
$$h_1 = y_1 = 0.8y_0 + 2x_1 = 0.8 \cdot 2 + 2 \cdot 0 = 1.6$$

Hence, the **impulse response** function is  $h_k = 0.8^k \cdot 2$  for  $k > 0$  which represents a **causal** system.

# Example: Calculating the impulse response function

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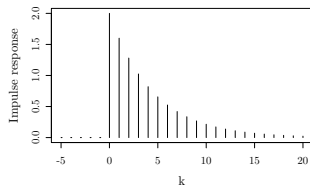
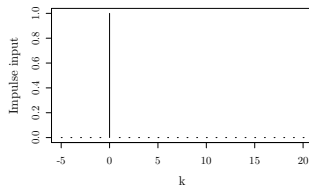
**Impulse response:** Send an impulse through the system:



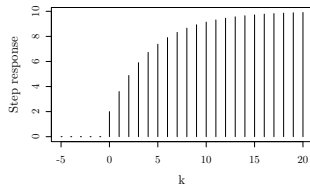
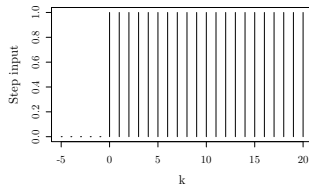
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$$\text{"ARX(1)": } y_t = 0.8y_{t-1} + 2x_t$$

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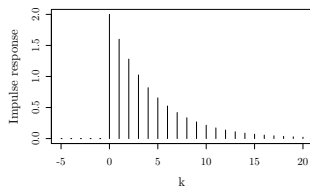
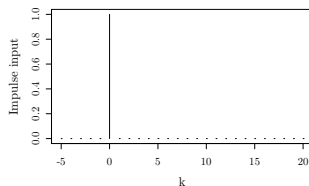
Is the system stable?

$$\sum_0^{\infty} |h_k| =$$

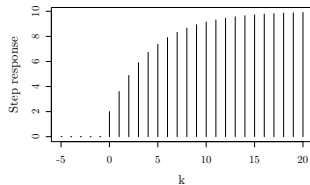
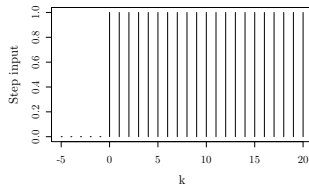
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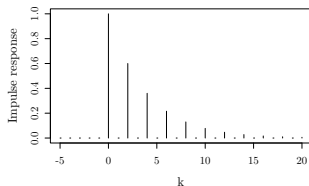
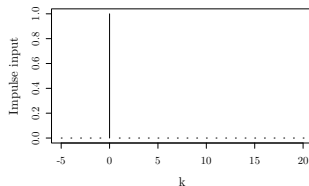


Is the system stable? YES, since  $\sum_0^{\infty} |h_k| = 10 < \infty$

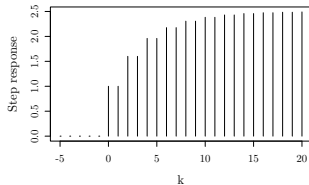
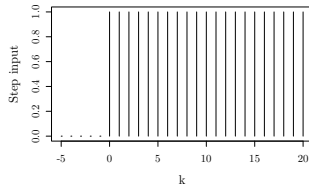
# Example: Calculating the impulse response function

$$\text{"ARX(2)": } y_t = 0.4y_{t-1} + 0.6y_{t-2} + x_t$$

**Impulse response:** Send an impulse through the system:



**Step response:** Send a step through the system!



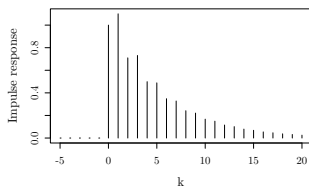
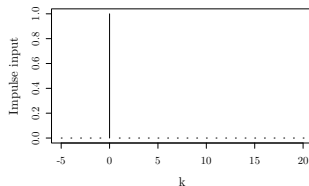
The system is **stable** since  $\sum_0^{\infty} |h_k| = \frac{10}{6} < \infty$



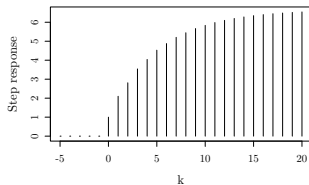
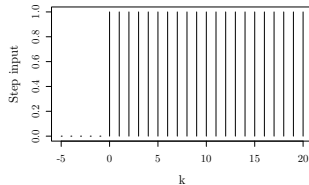
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"ARX(2)":  $y_t = 0.1y_{t-1} + 0.6y_{t-2} + x_t + x_{t-1}$

**Impulse response:** Send an impulse through the system:



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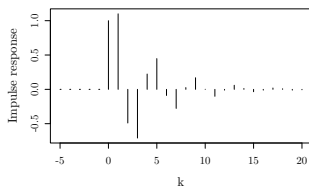
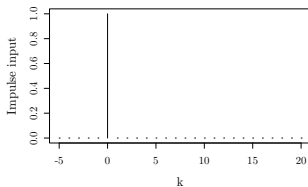


The system is **stable** since  $\sum_0^{\infty} |h_k| = \frac{20}{6} < \infty$

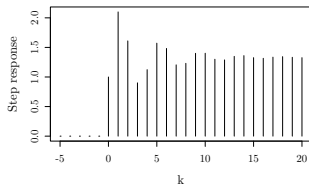
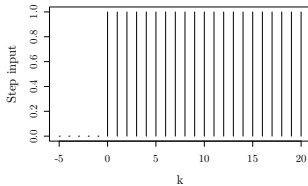
# Example: Calculating the impulse response function

"ARX(2)":  $y_t = 0.1y_{t-1} - 0.6y_{t-2} + x_t + x_{t-1}$

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**Step response:** Send a step through the system!

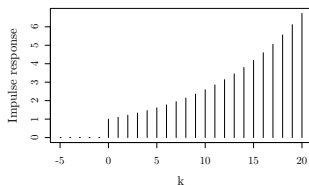
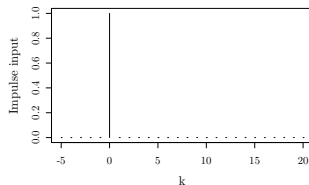


The system is **stable** since  $\sum_0^{\infty} |h_k| = \frac{4}{3} < \infty$

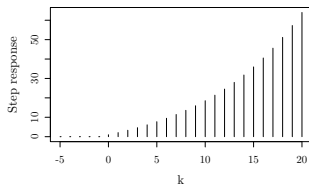
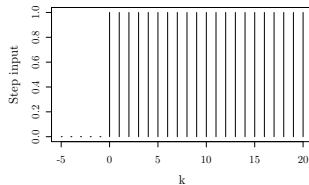
# Example: Calculating the impulse response function

"ARX(2)":  $y_t = 0.1y_{t-1} + 1.1y_{t-2} + x_t + x_{t-1}$

**Impulse response:** Send an impulse through the system:



**Step response:** Send a step through the system!



The system is **NOT stable** since  $\sum_0^{\infty} |h_k| = \infty$

Where do we try to observe the impulse or step response directly. Can you name some examples where it can be possible and useful to do so?

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- ▶ Sound: Clap or gun shot or blow-up balloon
- ▶ Exercise: From some day start doing regular exercise
- ▶ Experiment: Make a step increase in a set-point
- ▶ Biking: Letting go your hands from the bar, observe how you instinctive make a sort of impulse with your but to learn the response!
- ▶ Almost any activity where the system has some dynamics

## Dynamic response characteristics from data

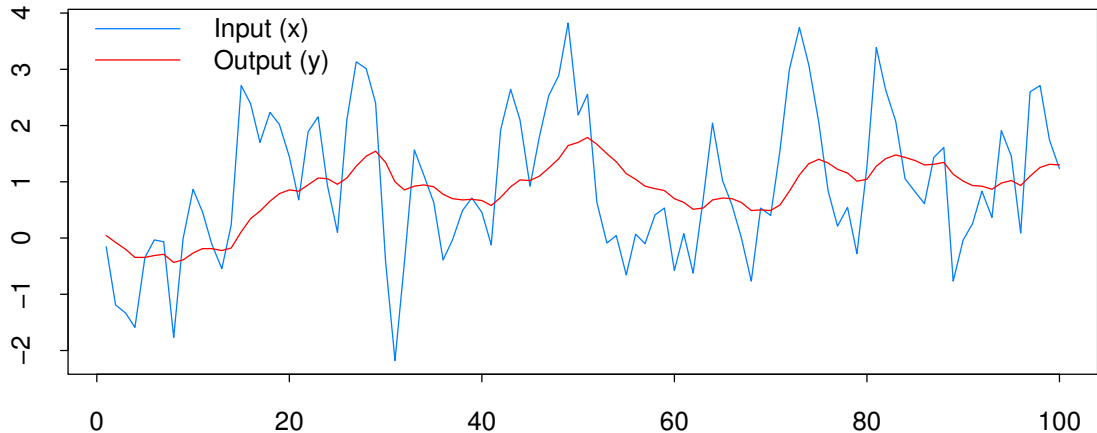
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# Stability based on the impulse response function

If the impulse response function is absolutely convergent, the system is stable (Theorem 4.3).

- ▶ Continuous time:

$$\int_{-\infty}^{\infty} |h(u)| du < \infty$$

- ▶ Discrete time:

$$\sum_{k=-\infty}^{\infty} |h_k| < \infty$$

# The $z$ -transform

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- ▶  $z = Ae^{i(\omega t + \phi)}$ :  $A$  is the amplitude,  $\omega$  is the angular frequency,  $t$  is time, and  $\phi$  is the phase. The frequency  $f$  can be derived from the angular frequency  $\omega$  using the relationship:

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$$y_t = \sum_{k=-\infty}^{\infty} h_k x_{t-k} \Leftrightarrow Y(z) = H(z)X(z)$$

*Time domain*  $\Leftrightarrow$  *Frequency domain*

# Linear Difference Equation

$$y_t + a_1 y_{t-1} + \cdots + a_p y_{t-p} = b_0 x_{t-\tau} + b_1 x_{t-\tau-1} + \cdots + b_q x_{t-\tau-q}$$
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Where the roots  $n_1, n_2, \dots, n_q$  are called the *zeros of the system* and  $\lambda_1, \lambda_2, \dots, \lambda_p$  are called the *poles of the system*.



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The system is stable if all poles lie within the unit circle

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The system is invertible if all zeroes lie within the unit circle

*In the course, we don't move longer into the frequency domain!*

## Estimating the impulse response

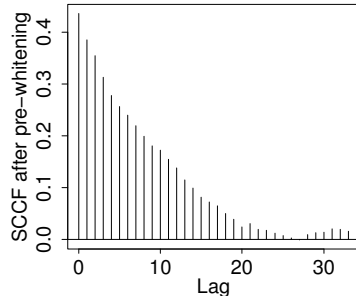
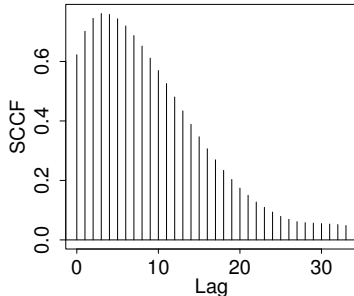
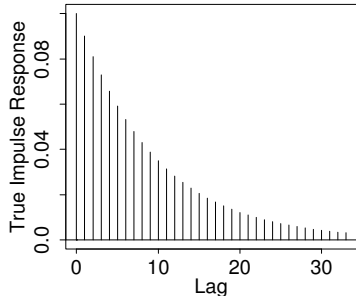
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- ▶ Alternative is to simply make an **LS with multiple inputs and all lags** (up to some max lag)! Works for FIR and ARX models.

# Estimating the impulse response

## ► Pre-whitening:

Identify a suitable ARMA to the input: Filter both input and output with the ARMA, and on residuals CCF is impulse response estimate.

**Pros:** Identify structure of ARMA, i.e. also MA part

**Cons:** Only works on single input and requires some (manual) modelling decision while doing it

## ► LS-estimates:

$\mathbf{Y} = \mathbf{X}\boldsymbol{\theta} + \boldsymbol{\varepsilon}$ , where columns of  $\mathbf{X}$  are the lagged inputs (Equation (8.49) in the Book).

**Pros:** No (manual) modelling decisions and works for multiple input models

**Cons:** Can not alone be used for identifying an MA part.

## Example: CCF vs. LS estimate of the impulse response

```

# Generate an AR(1) process as input
n <- 500
x <- c(arima.sim(list(ar=c(0.7,0,0)), n))
plot(x)

# Make an output vector "filtered" by the system:  $y_t = 0.8 * y_{t-1} + x_t$ 
y <- filter(x,0.8,"recursive") + rnorm(n)

# Calculate the true impulse response and plot with CCF
k <- 0:20
par(mfrow=c(2,1))
plot(k, 0.8^k, type="h", main="TRUE IR")
ccf(y,x, xlim=c(0,max(k)))

# Make lags for LS estimation
library(onlineforecast)
D <- as.data.frame(y=y, lagdf(x, k))
# See the model and the estimated result
(frml <- paste0("y ~ 0+",paste0("k",k,collapse="+")))
fit <- lm(frml, D)
summary(fit)

# Finally, Plot the true and the LS estimated IR
plot(k, 0.8^k, type="h", main="TRUE IR")
plot(0:20, fit$coef, type="h", xlab="lag", ylab="Impulse response")

```



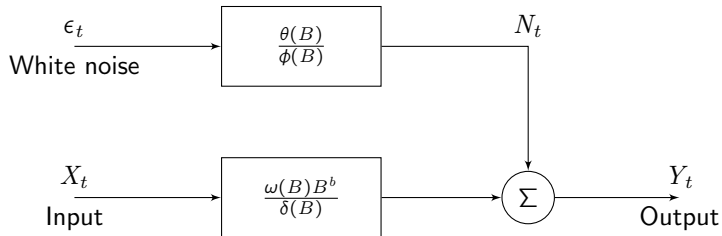
## Example: Loudspeaker and microphone system

Sound through loudspeaker and into mic example:



Open `impulse_reponse_record/Record.aup3`  
and then the analysis `example_IR_sound.R`.

# “Complete” Transfer function models



$$Y_t = \frac{\omega(B)}{\delta(B)} B^b X_t + \frac{\theta(B)}{\varphi(B)} \varepsilon_t$$

- ▶ Also called Box-Jenkins models

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- ▶ OE: Output Error model:  $Y_t = \frac{\omega(B)}{\delta(B)}B^b X_t + \epsilon_t$ .

- ▶ Regression models with ARMA noise (the `xreg` option to `arima` in R). Parameters are estimated in the same optimization:

$$(Y_t - \beta_0 + \beta_1 X_t) = (Y_t - \beta_0 + \beta_1 X_t) + \frac{\theta(B)}{\varphi(B)}\epsilon_t$$



# Identification of transfer function models

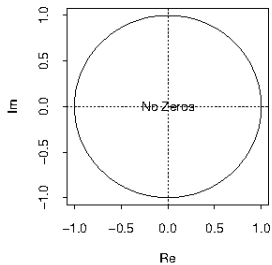
$$h(B) = \frac{\omega(B)B^b}{\delta(B)} = h_0 + h_1B + h_2B^2 + h_3B^3 + h_4B^4 + \dots$$

- ▶ Estimate the impulse response (pre-whitening or LS-estimate) and “guess” an appropriate structure of  $h(B)$  based on this.

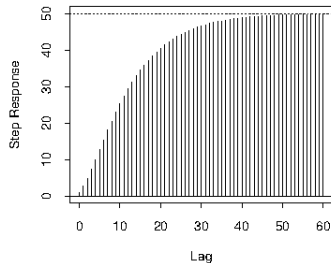
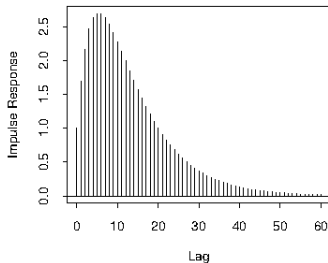
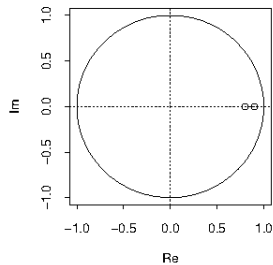
## 2 real poles

$$h(B) = \frac{1}{1 - 1.7B + 0.72B^2}$$

Zeros



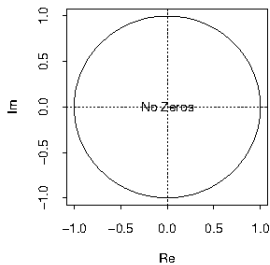
Poles



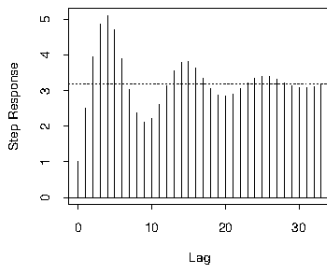
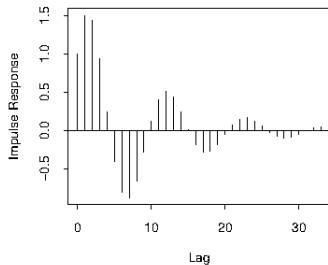
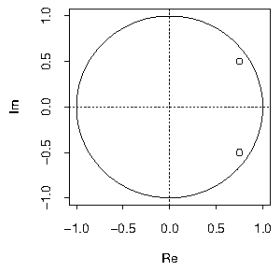
## 2 complex

$$h(B) = \frac{1}{1 - 1.5B + 0.81B^2}$$

Zeros



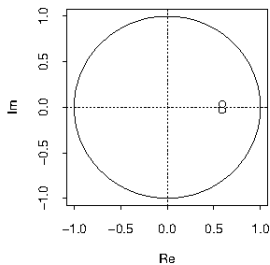
Poles



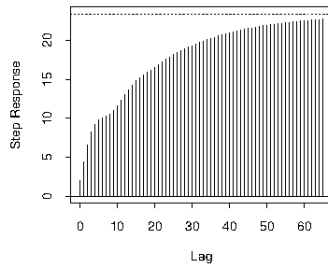
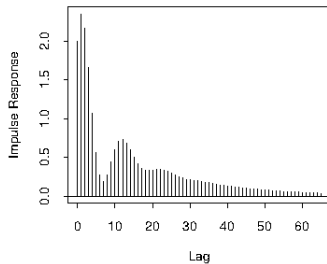
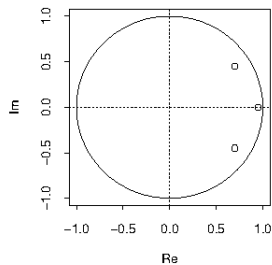
1 real, 2 comp

$$h(B) = \frac{2 - 2.35B + 0.69B^2}{1 - 2.35B + 2.02B^2 - 0.66B^3}$$

Zeros



Poles



# Identification of the transfer function for the noise

EITHER:

- ▶ After selection of the structure of the transfer function of the input we estimate the parameters of the model (assuming  $N_t$  to be white)

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OR

- ▶ Simply use a **forward or backward selection procedure**! It's often easier!



# Estimation

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- ▶ If model has MA-part (i.e. lagged residuals) some recursive method is needed (Kalman filter or Spliid method)
- ▶ For FIR and ARX models we can write the model as  $\mathbf{Y}_t = \mathbf{X}_t^T \boldsymbol{\theta} + \epsilon_t$  and use LS-estimates

# Model validation

As for ARMA models with some additions:

- ▶ Usual ACF of residuals and plots!
- ▶ Test for cross correlation between the residuals and the input. If  $\{\varepsilon_t\}$  is white noise and when there is no correlation between the input and the residuals then (approximately)

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- ▶ A *Portmanteau test* (*Ljung-Box*) can also be performed to test for significant ccf's.

# Cross covariance and cross correlation functions

Estimate of the cross covariance function:

$$C_{XY}(k) = \frac{1}{N} \sum_{t=1}^{N-k} (X_t - \bar{X})(Y_{t+k} - \bar{Y})$$
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What is a defining property of the CCF for causal systems with no feedback? If at least one of the processes is white noise and if the processes are uncorrelated then  $\hat{\rho}_{XY}(k)$  is approximately normally distributed with mean 0 and variance  $1/N$